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**IEEE P802.11  
Wireless LANs**

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**An Improved Rate Signalling**

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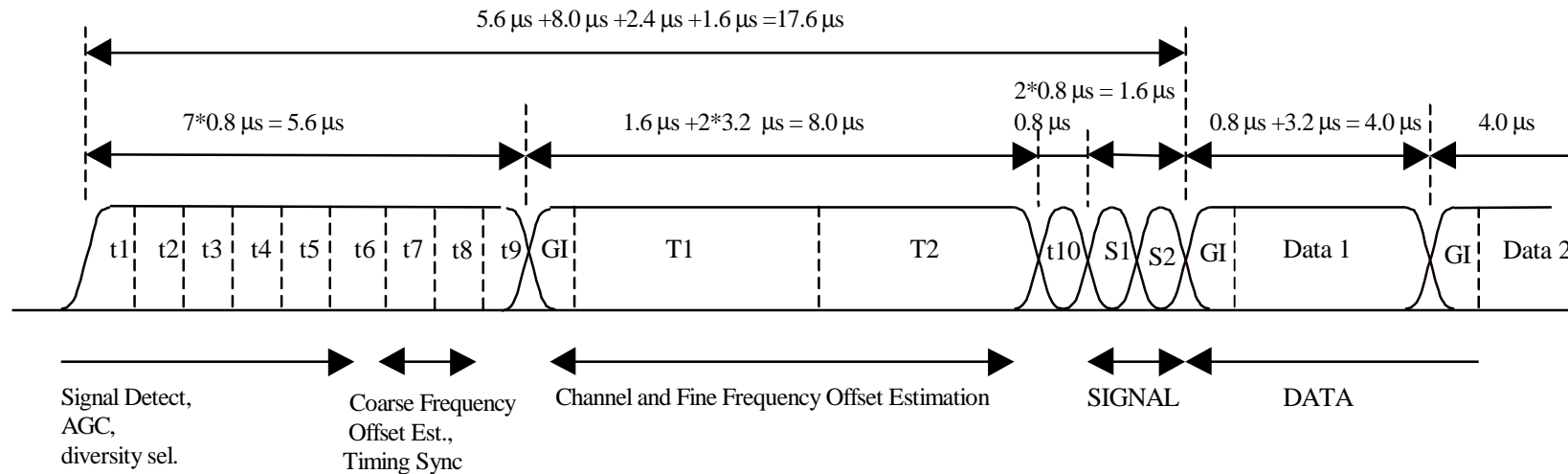
**Author:** Naftali Chayat and Tal Kaitz  
BreezeCom  
Atidim Technology Park, Tel Aviv 61131 Israel  
Phone: 972 -3-6456262  
Fax: 972-3-6546290  
e-Mail: {Talk, Naftalic}@ Breezecom.co.il

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**An Improved Rate Signalling**

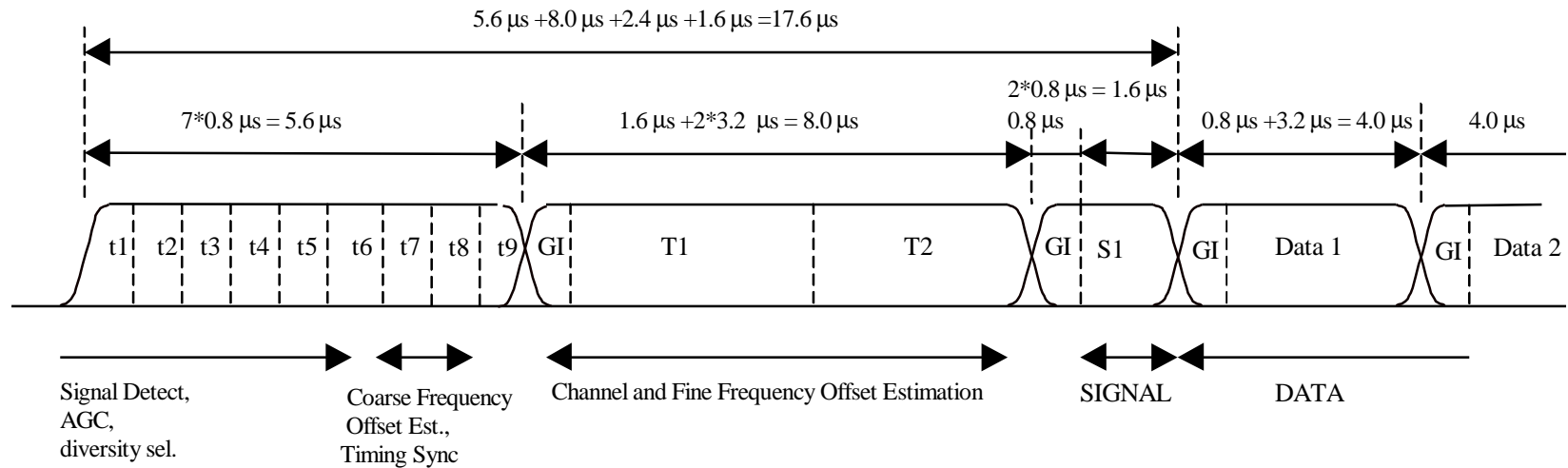
- The reliability of the rate-signaling scheme is crucial for the performance of the 802.11a Wireless LAN.
- We propose a new reliable scheme based on bi-orthogonal Hadamard coding and OFDM modulation.
- No overhead relative to current scheme
- Simple to implement.

## Overview of current scheme



- 4 bits are conveyed by QPSK modulating the sequences S1 and S2.
- t10 serves as a guard interval.
- Overall length  $0.8 \mu\text{s} + 0.8 \mu\text{s} + 0.8 \mu\text{s} = 2.4 \mu\text{s}$ .
- In AWGN: same error rate as rate  $\frac{1}{2}$  BPSK OFDM. (6Mb/s).
- Non satisfactory performance under severe multipath conditions.

## Proposed Scheme



- 4 bits are conveyed by the sequence S1 spanning 1.6uS.
- Additional guard interval of 0.8uS.
- Overall length 2.4uS – same as before.

## Coding and Modulation

- 3 LSB select one row of Hadamard 8 matrix.
- The MSB selects sign.
- The 8 binary symbols are repeated 3 times to form 24 vector.
- The vector is multiplied with a cover sequence.
- The result is used to modulate the even subcarriers of a 64 point OFDM symbol. The time domain vector has two identical halves.
- The time domain vector is cyclically extended and a window is applied to truncate it to length  $2.4uS$ .

## Decoding and Demodulation

- The 32 samples signal is cyclically extended to provide 64 samples.
  - A 64-point FFT is used to demodulate.
  - The even subcarriers are multiplied by the cover sequence.
  - The subcarriers are combined to produce an 8-point vector.
  - A Fast Hadamard Transform is applied.
  - The location of peak determines 3 MSBs. The sign of the peak determines the MSB.
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- Both modulation and demodulation require existing H/W: namely the 64 point FFT/IFFT
  - Coding and decoding require an 8 point fast Hadamard transform.

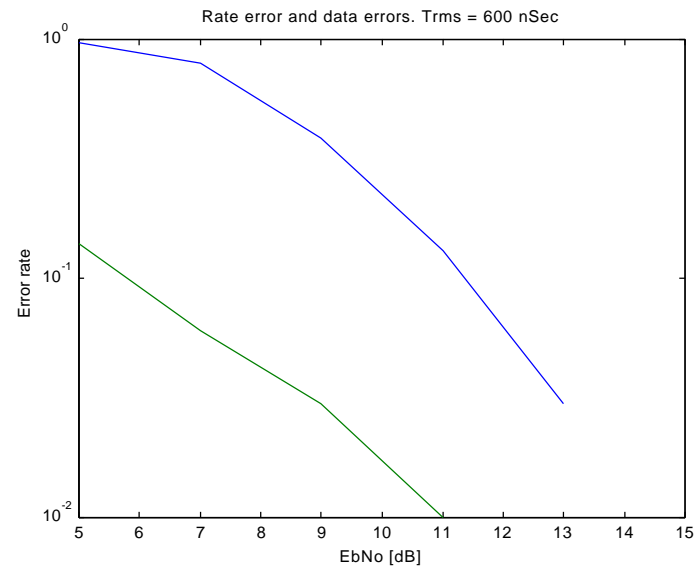
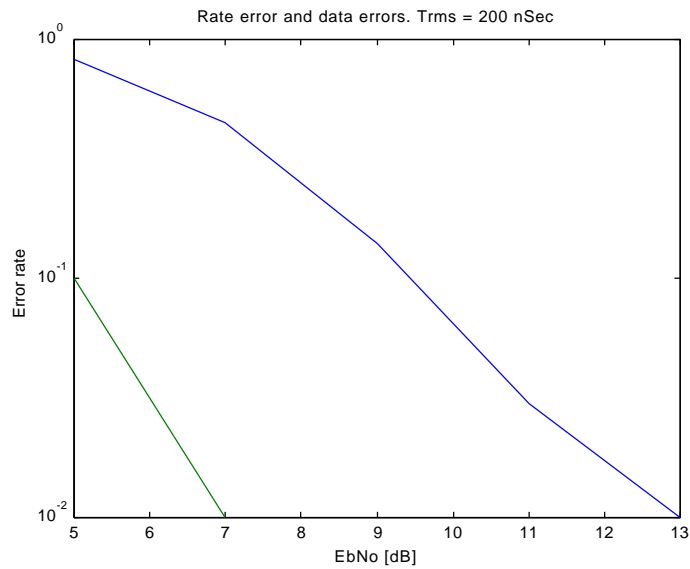
### Performance in flat AWGN

- Let  $E_s$  denote the power per spectral line.
  - Then for rate  $\frac{1}{2}$   $k=7$  BPSK we have  $d_{\text{free}}^2 = 10 * 4E_s = 40E_s$ .
  - For bi-orthogonal coding we have  $d_{\text{free}}^2 = 2 * 8 * 3 * E_s = 48E_s$ .
- ⇒ The bi-orthogonal scheme is better by 0.8dB than the coding scheme of the data section.

## Performance in severe multipath

Simulation results: 64bytes packets, 6Mb/s.

Green : errors in rate field. Blue: errors in data.



## Peak to Average Power Ratio

The cover sequence assures good PAP ratios for all codewords.

| Codeword | PAP [dB] |
|----------|----------|
| 1        | 3.2 dB   |
| 2        | 3.7 dB   |
| 3        | 4.6 dB   |
| 4        | 3.7 dB   |
| 5        | 3.2 dB   |
| 6        | 4.2 dB   |
| 7        | 4.6 dB   |
| 8        | 3.7 dB   |

PAP = 3.2dB... 4.6dB.



## Extension to 5 bits

- Due to the proliferation of codes and data rates, (1/3 , 9/16 etc.) there is a need to convey more than 4 bits.
- The proposed scheme can be easily extended to support 5 bits.
- Performed by QPSK modulating the  $H_8$  row.
- Decoding by complex 8-point FHT.
- Same minimum free distance of  $48 \cdot E_s$ . Double number of minimum distance neighbours => Slight degradation in performance.

## Conclusions

- A reliable method for transmitting the 4 bits of the rate field.
  - Requires no overhead relative to the current scheme.
  - Much lower error rate than the data section even in the most reliable mode, both under flat channel and under severe multipath conditions.
  - Simple to implement. Uses existing modulation and demodulation mechanisms.
  - Requires Fast Hadamard Transform to be implemented.
- Simple extension to 5 bits.